

MINT-MERIT

**Master of Science in Information and Communication
Technologies**

COMMUNICATION THEORY

Continuous Time/Amplitude Channels

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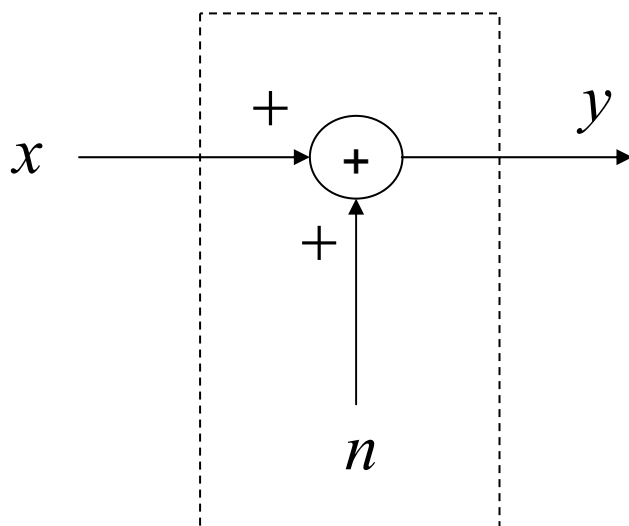
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THE ADDITIVE (WHITE) GAUSSIAN CHANNEL: A(W)GC

We will now focus on the following important channel model:



$$x \sim f_x(x)$$

$$n \sim N(0, \sigma_n^2) \quad \text{with } f_n(n)$$

If 'x' and 'n' independent R.V.:

$$f_y(y) = f_x(\lambda) * f_n(\lambda) \Big|_{\lambda=y}$$



(‘EXTENSION’ OF THE INFORMATION DEFINITION FOR THE CONTINUOUS DISTRIBUTIONS)

Let’s X to be a R.V. with probability density function (p.d.f.) $f_x(X)$

We define entropy $H(X)$ as follows (Differential Entropy)

$$H(X) \equiv \int_{-\infty}^{+\infty} f_x(x) \log \frac{1}{f_x(x)} dx \geq 0!! \quad \text{and/or arbitrary large}$$

For two R.V. ‘x’ and ‘y’, we also define:

$$H(x; y) = \iint f(x; y) \log \frac{1}{f(x; y)} dx dy$$

$$H(x / y) = \iint f(x; y) \log \frac{1}{f(x / y)} dx dy \quad \text{and } H(y/x) = \dots$$



When both $H(x)$ and $H(y)$ are finite, then:

$$H(x; y) \leq H(X) + H(Y)$$

$$H(x; y) = H(X) + H(Y / X) = H(Y) + H(X / Y)$$

$$H(Y / X) \leq H(Y)$$

$$H(X / Y) \leq H(X)$$

All them become equalities (“=”) for ‘x’ and ‘y’ independent.



THEOREM:

For a R.V. with p.d.f. given by $f_x(x)$ and finite variance σ_x^2 , the entropy (differential) is upper-bounded by:

$$H(X) \leq \frac{1}{2} \log(2\pi e \sigma_x^2)$$

with equality iff 'x' follows a gaussian distribution $x \sim N(\mu; \sigma_x^2)$



This problem follows from the variational solution to:

$$f_x(x) \equiv \arg \max_{f_x(x)} \int_{-\infty}^{+\infty} f_x(x) \log \frac{1}{f_x(x)} dx$$

Subject to:

$$\int_{-\infty}^{+\infty} (x - \mu)^2 f_x(x) dx = \sigma_x^2$$

and:

$$\int_{-\infty}^{+\infty} f_x(x) dx = 1 \quad f_x(x) \geq 0$$

It is also possible to proceed as follows. We first state that:

$$+ \int_{-\infty}^{+\infty} f_x(x) \log \frac{f_x(x)}{g(x)} dx \geq 0; \quad \text{for any p.d.f. } g(x)$$

$$" = " \quad \underline{\text{iif}} \quad f_x(x) = g(x) \quad \forall x$$

We see that:

$$\int_{-\infty}^{+\infty} f_x(x) \log \frac{g(x)}{f_x(x)} dx \stackrel{(*)}{\leq} \frac{1}{\ln 2} \int f_x(x) \left[\frac{g(x)}{f_x(x)} - 1 \right] dx = 0$$

Thus:

$$H(X) \leq \int_{-\infty}^{+\infty} f_x(x) \log \frac{1}{g(x)} dx \quad " = " \quad \underline{\text{iif}} \quad f_x(x) = g(x)$$

We assume 'x' and 'y' two R.V. with the same mean 'μ' and variance σ², such that f_x(x) is the p.d.f. for 'x' and 'y' is a gaussian R.V. y~N(μ,σ²):

We see that:

$$\begin{aligned}
 H[Y] &= \int f_y(y) \log \frac{1}{f_y(y)} dy = \int f_y(y) \left(\frac{1}{2} \log(2\pi\sigma^2) + \frac{(y-\mu)^2}{2\sigma^2} \right) dy = \\
 &= \frac{1}{2} \log(2\pi\sigma^2) + \frac{1}{2\sigma^2} E[(y-\mu)^2] = \frac{1}{2} \log(2\pi\sigma^2) + \frac{1}{2\sigma^2} E[(x-\mu)^2] = \\
 &= \int f_x(x) \log \frac{1}{f_y(x)} dx \geq \int f_x(x) \log \frac{1}{f_x(x)} dx = H[X]
 \end{aligned}$$

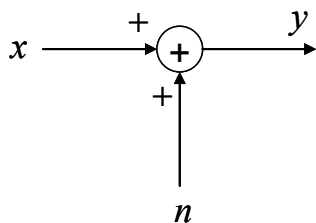
"=" iff $f_x(x) = f_y(x) \quad \forall x$

Then, for:

$$g(x) = \frac{1}{\sqrt{2\pi\sigma_x}} e^{-\frac{1}{2}\left(\frac{x-\mu}{\sigma_x}\right)^2}$$

$$\begin{aligned} H(X) &\leq \int_{-\infty}^{+\infty} f_x(x) \log \frac{1}{g(x)} dx = \\ &= \frac{1}{2} \log(2\pi\sigma_x^2) + \frac{\log e}{2\sigma_x^2} \underbrace{\int_{-\infty}^{+\infty} (x-\mu)^2 f_x(x) dx}_{\sigma_x^2} = \\ &= \frac{1}{2} \log(2\pi e \sigma_x^2) !! \quad " = " \quad \text{iif } f_x(x) = g(x) \forall x \end{aligned}$$

CAPACITY FOR A GAUSSIAN CHANNEL



$$n \sim N(0, \sigma_n^2)$$

$$I(A_x; A_y) = H(Y) - H(Y / X)$$

- For a given 'x':

$$y \sim N(x, \sigma_n^2) \Rightarrow H(Y / X) = \frac{1}{2} \log(2\pi e \sigma_n^2)$$

For $H(Y)$, we have that:

$$y = x + n \quad \text{with 'x'; 'n' independent R.V.}$$

thus: $\sigma_y^2 = \sigma_x^2 + \sigma_n^2$

Finally:

$$C \equiv \max_{f_x(x)} I(A_x; A_y) \Rightarrow \text{if } \left. \begin{array}{l} x \sim N(\mu; \sigma_x^2) \\ n \sim N(0; \sigma_n^2) \end{array} \right| \Rightarrow y = x + n \sim N(\mu; \sigma_x^2 + \sigma_n^2)$$

$$C = \underbrace{\frac{1}{2} \log(2\pi e (\sigma_x^2 + \sigma_n^2))}_{H(Y)} - \underbrace{\frac{1}{2} \log(2\pi e \sigma_n^2)}_{H(Y/X)} = \frac{1}{2} \log \left(1 + \frac{\sigma_x^2}{\sigma_n^2} \right) \quad !!!$$

Thus, for a discrete-time gaussian channel:

$$C = \frac{1}{2} \log \left(1 + \frac{\sigma_x^2}{\sigma_n^2} \right) = \frac{1}{2} \log(1 + SNR) \quad \text{with : } SNR \equiv \frac{\sigma_x^2}{\sigma_n^2}$$

If the channel is a band-limited channel in the interval $(-BW, +BW)$ and for an additive WHITE gaussian noise (AWGN):

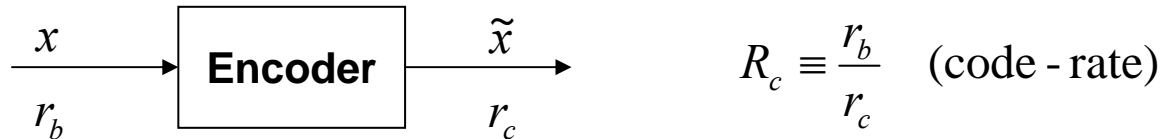
$$\sigma_n^2 = \frac{N_o}{2} 2BW = N_o BW$$
$$C = \frac{1}{2} \log \left(1 + \frac{\sigma_x^2}{N_o BW} \right)$$
$$\left(S_{nn}(f) = \frac{N_o}{2} \text{ watts / Hz} \right)$$

\uparrow
($e^{j2\pi f}$)

For a transmission rate r_b bits/sec, we can also define the “*Energy per Information Bit*” or E_b as follows:

$$E_b \equiv \sigma_x^2 T_b \equiv \frac{\sigma_x^2}{r_b} \quad \Rightarrow \quad C = \frac{1}{2} \log \left(1 + \frac{E_b r_b}{N_o BW} \right)$$

Impact of the encoder:



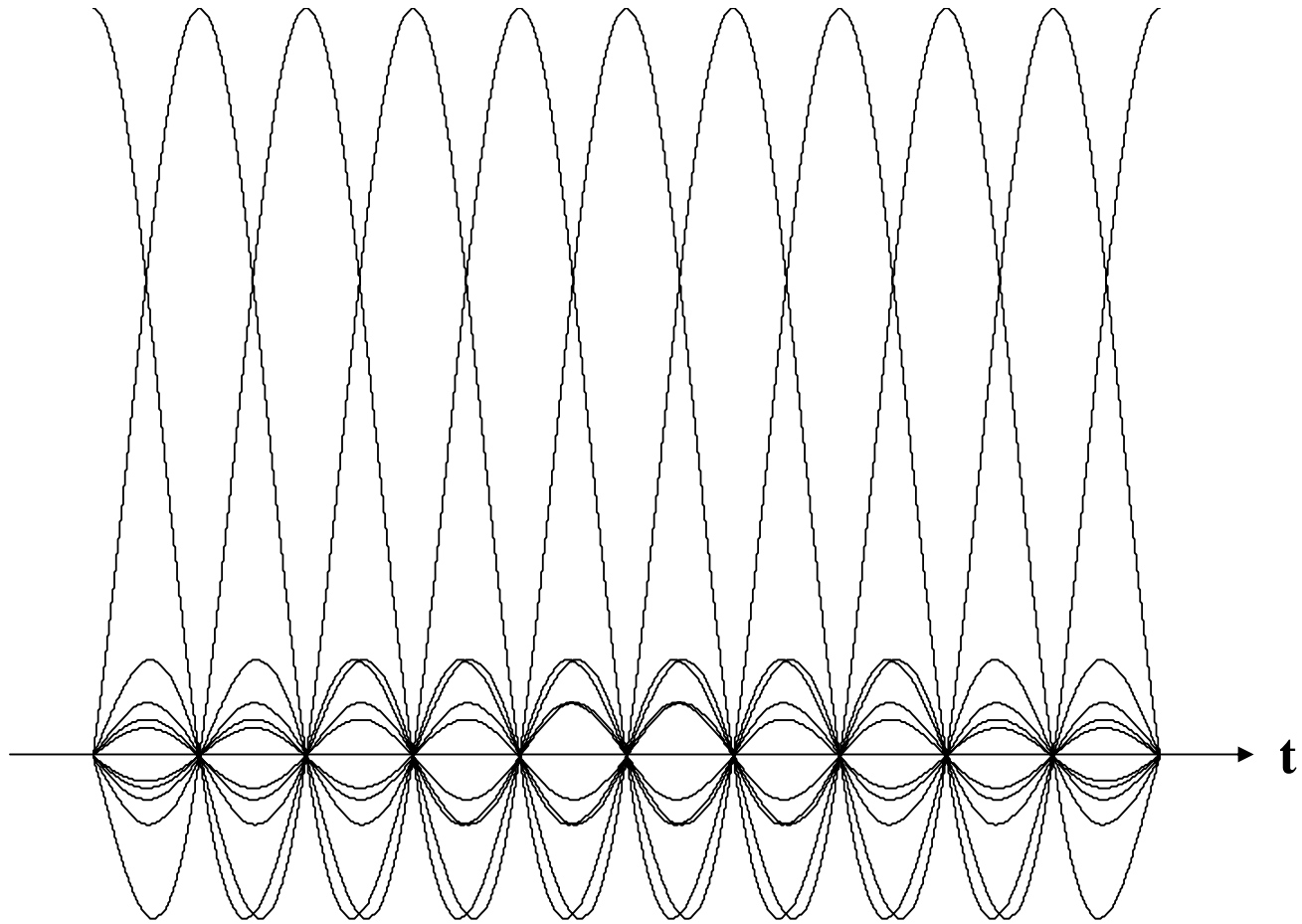
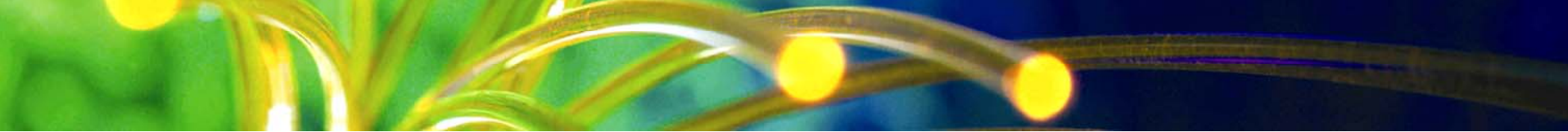
Minimum transmission bandwidth (Nyquist Bandwidth): $BW = \frac{r_c}{2} = \frac{r_b}{2R_c}$

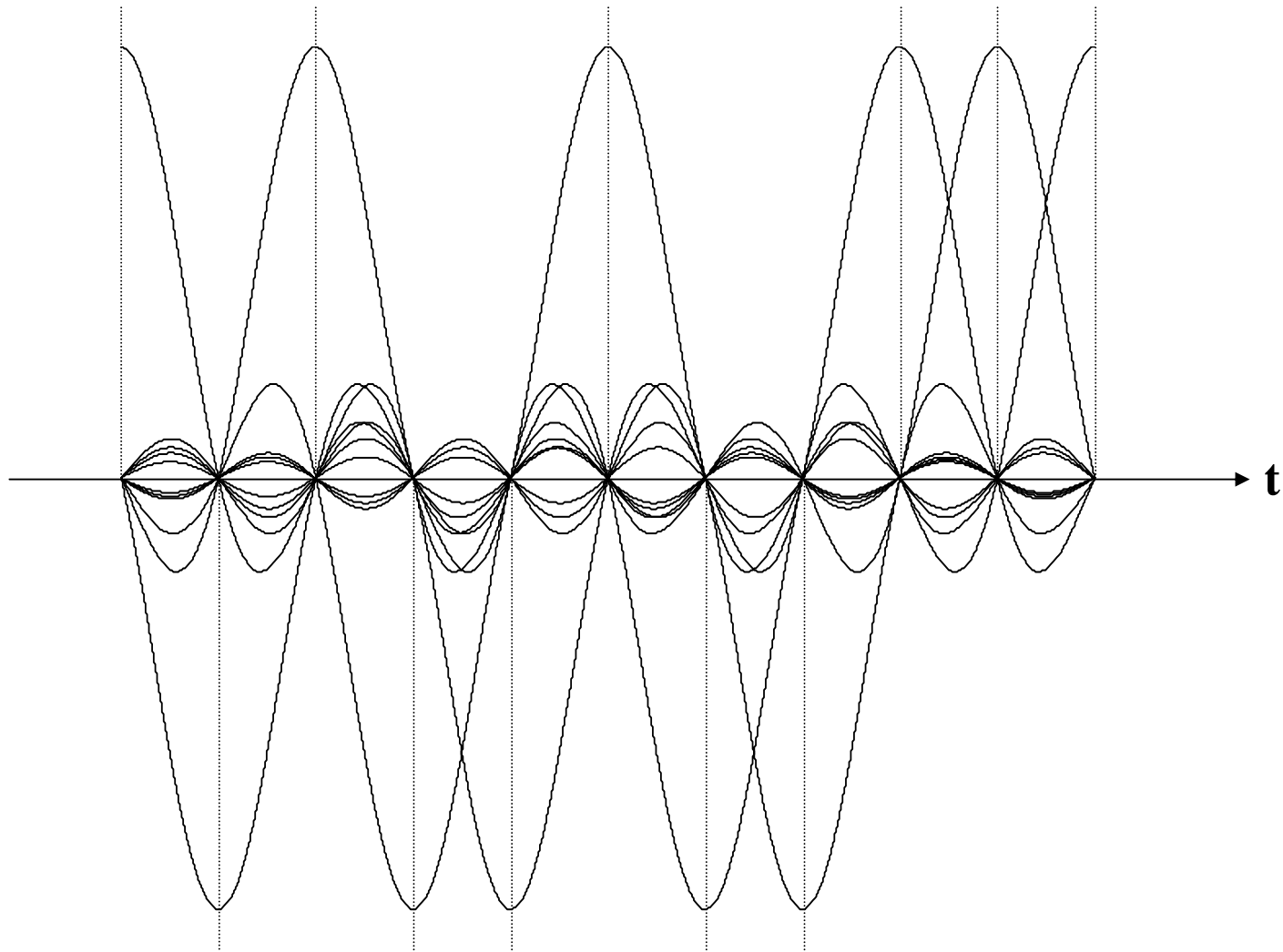
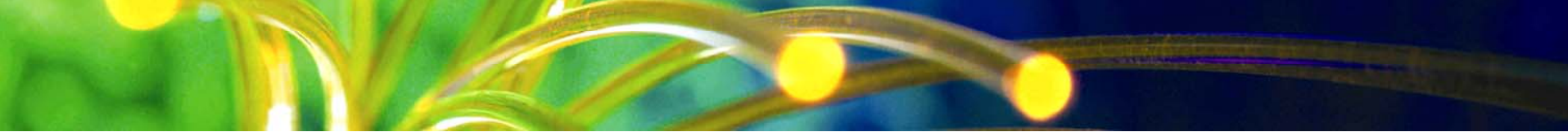
Thus: $C = \frac{1}{2} \log \left(1 + 2R_c \frac{E_b}{N_o} \right)$ bits/channel use

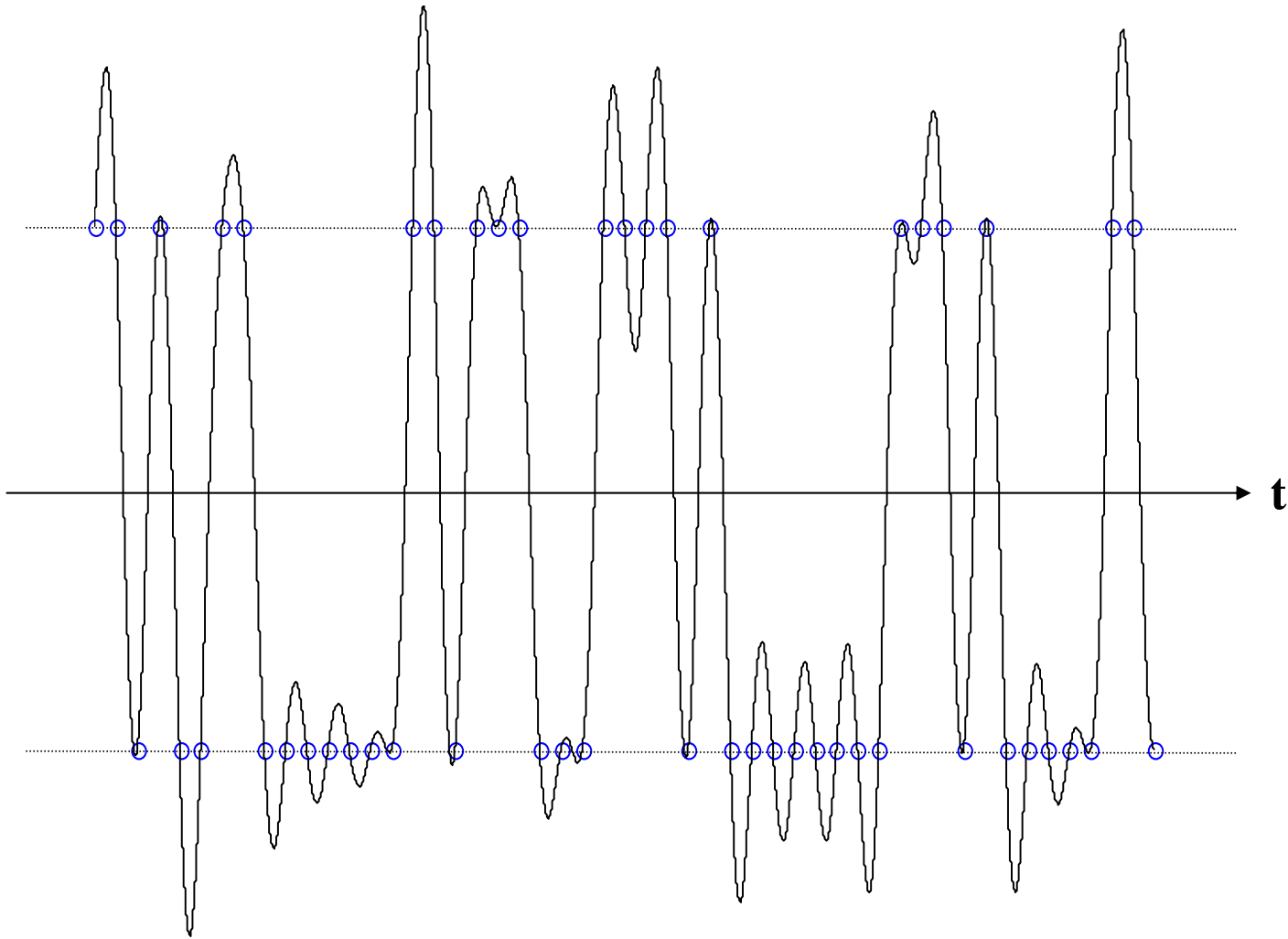
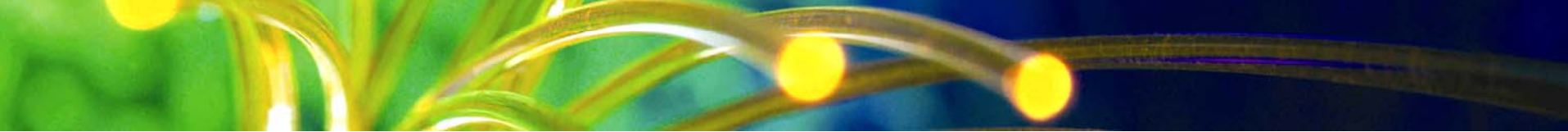
For a channel with bandwidth BW (Hz), a total of 2BW channel-uses/sec are possible still preserving the DMC nature of the channel (!):

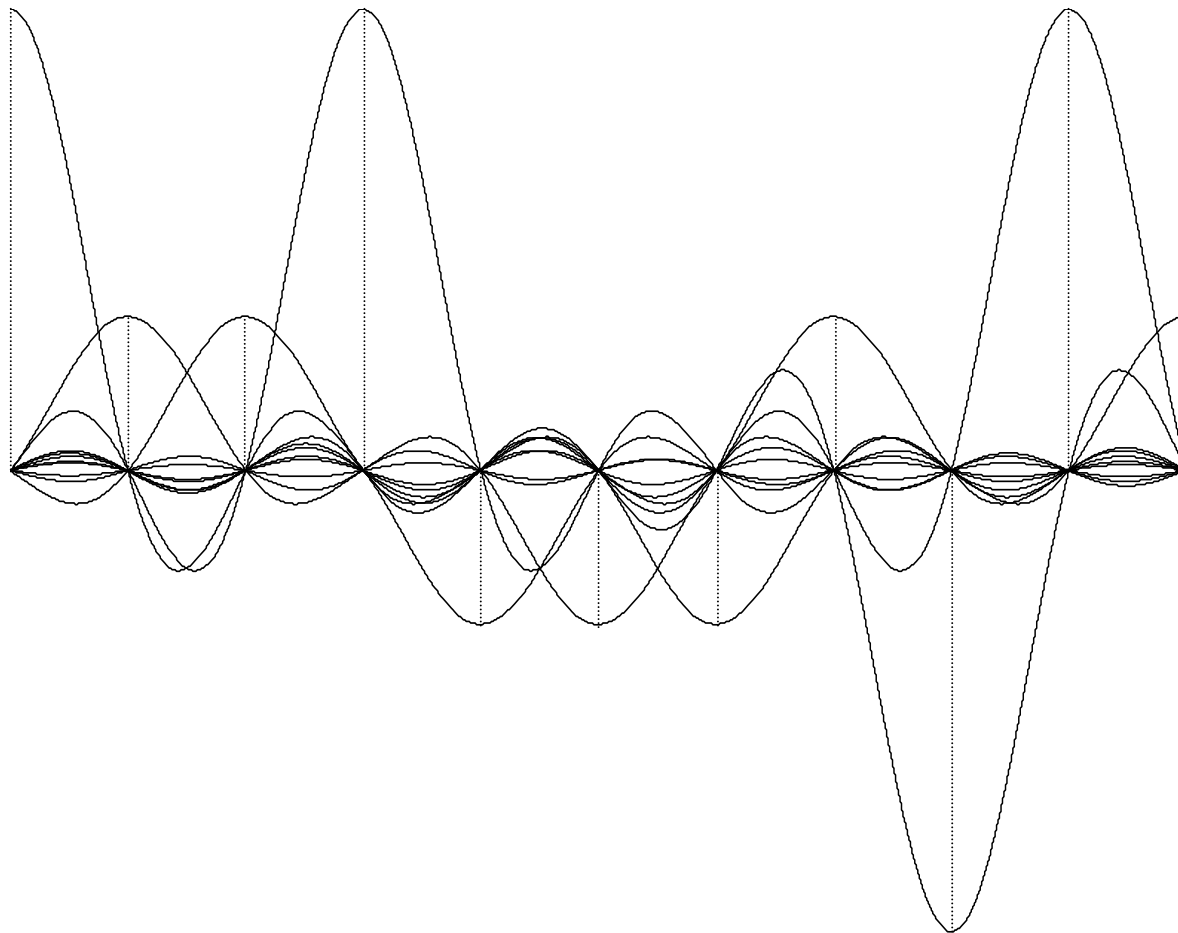
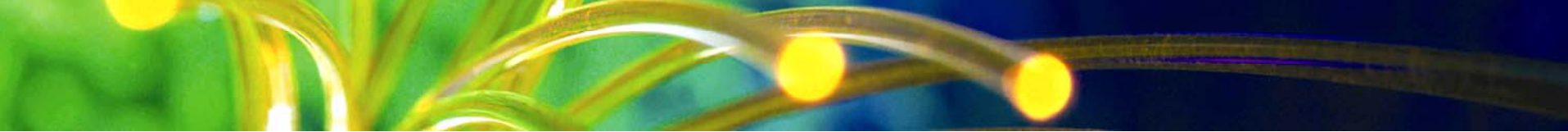
$$\begin{aligned} C &= BW \log(1 + SNR) \stackrel{(*)}{=} BW \log \left(1 + \frac{\sigma_x^2}{N_o BW} \right) = BW \log \left(1 + \frac{E_b r_b}{N_o BW} \right) = \\ &= BW \log \left(1 + 2R_c \frac{E_b}{N_o} \right) \text{ bits/sec} \end{aligned} \quad (*) \text{ White noise}$$

(!) Moreover, 2BW samples/sec suffices for a perfect time representation of a band-limited signal of bandwidth BW (Hz).









Finally, for an efficient code ($R_c=1$) we can relax the bandwidth constraint:

$$\lim_{BW \rightarrow \infty} C \longrightarrow \lim_{BW \rightarrow \infty} \log \left(1 + \frac{E_b r_b}{N_o BW} \right)^{BW} = \lim_{BW \rightarrow \infty} \log \left(1 + \frac{1}{\underbrace{BWN_o / E_b r_b}_e} \right)^{\frac{BWN_o}{E_b r_b} \cdot \frac{E_b r_b}{N_o}} \longrightarrow$$

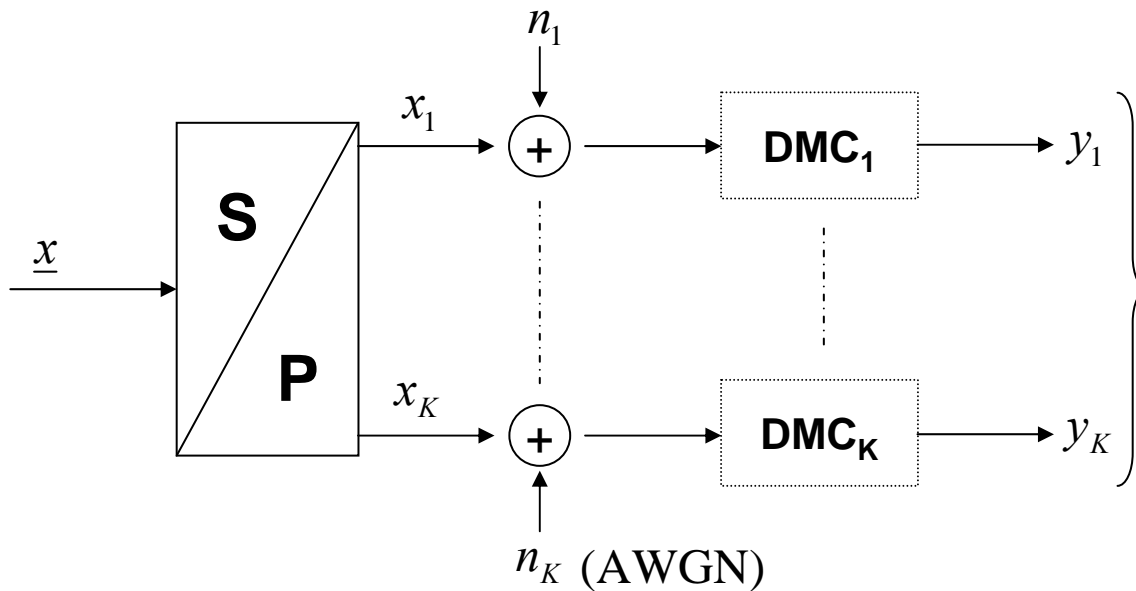
$$\longrightarrow \frac{E_b r_b}{N_o} \log e$$

$$\boxed{\lim_{BW \rightarrow \infty} C \rightarrow \frac{E_b r_b}{N_o} \log e} \text{ bits/sec}$$

Thus:

$$r_b \leq C \Rightarrow r_b \leq \frac{E_b r_b}{N_o} \log e \Rightarrow \frac{E_b}{N_o} \geq \frac{1}{\log e} \cong \underline{\underline{-1.59dB!!}}$$

(Spectral Efficiency (E_r) is defined as : $E_r \equiv \frac{r_b}{BW}$ bits/sec/Hz)



$$C_i = BW_i \log(1 + SNR_i)$$

$$\text{with: } SNR_i \equiv \frac{\sigma_{x_i}^2}{\sigma_{n_i}^2}$$

Because the independence + the DMC condition:

$$H(\underline{Y}) = \sum_i H(Y_i)$$

$$H(\underline{Y} / \underline{X}) = \sum_i H(Y_i / X_i)$$

$$C = \sum_i \underbrace{BW_i \log(1 + SNR_i)}_{C_i} = \log \prod_{i=1}^K (1 + SNR_i)^{BW_i} \dots$$

(Consequences!)

THEOREM:

The capacity for 'K' parallel additive gaussian channels under a total transmitted power S_T constraint is given by:

With:

$$C(S_T) = \sum_{k=1}^K \frac{1}{2} \log \left(1 + \frac{\sigma_i^2}{\sigma_{ni}^2} \right)$$



$$\left| \begin{array}{l} \Rightarrow \sigma_i^2 = \max(0; \lambda - \sigma_i^2) \\ \Rightarrow ' \lambda ' \text{ such that } \sum_{k=1}^K \sigma_i^2 = S_T \end{array} \right. \longrightarrow \text{“WATER-POURING” concept}$$

This capacity is achieved if the channels are gaussian-independent.

1.- We saw that:

$$I(\underline{\mathbf{x}}; \underline{\mathbf{y}}) \leq \sum_{k=1}^K I(x_k; y_k) \quad " = " \text{ if independent (sufficient)}$$



$$\frac{1}{2} \log \left(1 + \frac{\sigma_i^2}{\sigma_{ni}^2} \right) \quad i = 1, 2, \dots, K$$

2.- From the Lagrange Multiplier technique:

$$\left. \begin{array}{l} \max_{\{\sigma_k^2\}} \sum_{k=1}^K \frac{1}{2} \log \left(1 + \frac{\sigma_k^2}{\sigma_{nk}^2} \right) \\ \text{Subject to:} \\ \sum_{k=1}^K \sigma_k^2 = S_T \end{array} \right\} \equiv \max_{\{\sigma_k^2\}} \underbrace{\sum_{k=1}^K \frac{1}{2} \log \left(1 + \frac{\sigma_k^2}{\sigma_{nk}^2} \right)}_{\Phi[\underline{\sigma}^2]} + \mu \left[\sum_{k=1}^K \sigma_k^2 - S_T \right]$$

Thus:

$$\frac{\partial}{\partial \sigma_i^2} \Phi(\underline{\sigma}^2) = \frac{1}{2} \cdot \frac{1/\sigma_{n_i}^2}{1 + \sigma_i^2/\sigma_{n_i}^2} + \mu = \frac{1}{2} \cdot \frac{1}{\sigma_i^2 + \sigma_{n_i}^2} + \mu = 0$$

That is:

$$\sigma_i^2 = -\frac{1}{2\mu} - \sigma_{n_i}^2 \geq 0 !! \quad \Rightarrow \quad \boxed{\sigma_i^2 = \lambda - \sigma_{n_i}^2 \geq 0}$$

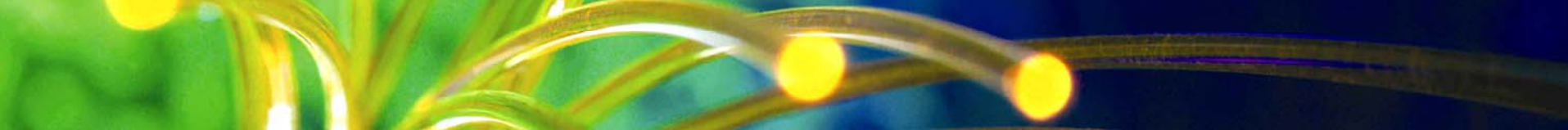
$$\text{For: } \lambda = -\frac{1}{2\mu}$$

$$\lambda = -\frac{1}{2\mu} = \left[S_T + \sum_{k=1}^{K_o} \sigma_{n_k}^2 \right] \frac{1}{K_o}$$

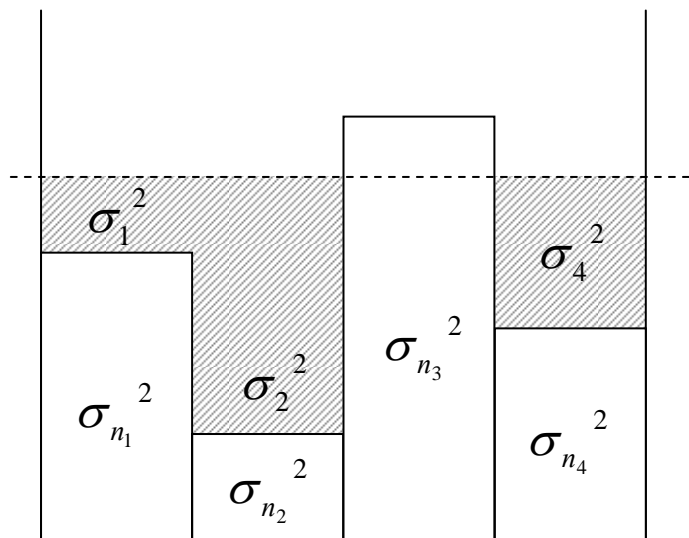
$$K_o = \sum_{n=1}^K K_n$$

$$K_n = \begin{cases} 1 & \text{if } \lambda - \sigma_{n_k}^2 > 0 \\ 0 & \text{if } \lambda - \sigma_{n_k}^2 \leq 0 \end{cases}$$

$$\boxed{\sigma_i^2 = 0 \quad \text{for } \lambda - \sigma_{n_i}^2 < 0}$$



Water-Pouring (Water-Filling):



$$S_T = \sigma_1^2 + \sigma_2^2 + \sigma_3^2 + \sigma_4^2$$

$$\sigma_i^2 = \max(0; \lambda - \sigma_{ni}^2)$$

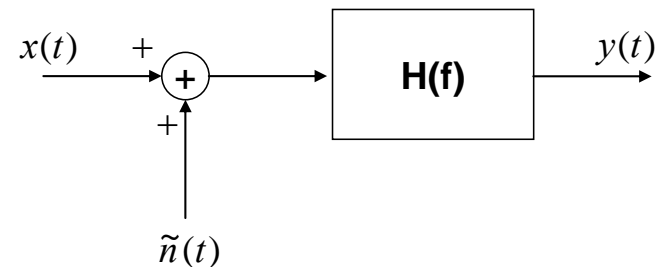
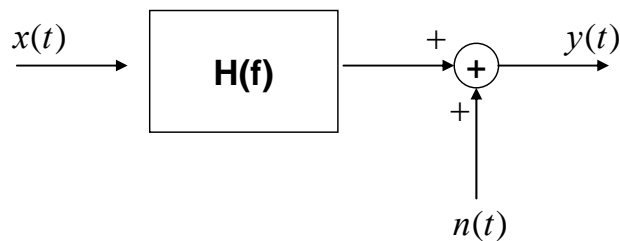
$$\lambda = \frac{1}{K_o} \left[S_T + \sum_{k=1}^{K_o} \sigma_{nk}^2 \right]$$

MORE GENERAL CASES:

1.- Define the “signal to noise spectral density ratio”

$$SNR(f) \equiv \frac{S_{xx}(f)}{S_{nn}(f)} \Rightarrow C = \int_{BW} \log \left(1 + \frac{S_{xx}(f)}{S_{nn}(f)} \right) df$$

2.- Frequency-selective channels (channels with memory)



(think on the impact of the channel memory...)

$$S_{\tilde{n}\tilde{n}}(f) = \frac{S_{nn}(f)}{|H(f)|^2}$$

- Wizard Setup
- Advanced Setup
- Maintenance**
 - System Status
 - DHCP Table
 - Diagnostic**
 - Firmware
- Logout

Diagnostic - DSL Line

```

relative capacity occupation: 44%
noise margin upstream: 26.0 db
output power downstream: 20.0 dbm
attenuation upstream: 15.0 db
carrier load: number of bits per symbol(tone)
tone 0- 31: 00 00 00 02 22 45 66 77 77 66 65 66 54 43 33 00
tone 32- 63: 00 00 00 00 00 00 00 22 22 22 22 33 33 44 44 44
tone 64- 95: 44 42 44 44 44 44 44 44 44 44 44 44 44 33 33 33
tone 96-127: 33 32 22 22 22 22 22 22 22 22 22 22 22 22 22 22
tone 128-159: 22 22 22 22 22 22 22 22 22 22 22 22 22 22 22 22
tone 160-191: 22 22 22 22 22 20 22 22 22 22 22 22 22 22 22 22
tone 192-223: 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00
tone 224-255: 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00

```

Back

- Wizard Setup
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Diagnostic - DSL Line

```

relative capacity occupation: 16%
noise margin downstream: 39.5 db
output power upstream: 12.0 dbm
attenuation downstream: 31.0 db
carrier load: number of bits per symbol(tone)
tone 0- 31: 00 00 00 02 22 45 66 77 77 66 65 66 54 43 33 00
tone 32- 63: 00 00 00 00 00 00 00 22 22 22 22 33 33 44 44 44
tone 64- 95: 44 42 44 44 44 44 44 44 44 44 44 44 44 33 33 33
tone 96-127: 33 32 22 22 22 22 22 22 22 22 22 22 22 22 22 22
tone 128-159: 22 22 22 22 22 22 22 22 22 22 22 22 22 22 22 22
tone 160-191: 22 22 22 22 22 20 22 22 22 22 22 22 22 22 22 22
tone 192-223: 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00
tone 224-255: 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00 00

```

Back

- ZyXEL**
TOTAL INTERNET ACCESS SOLUTION
- Wizard Setup
- Advanced Setup
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[SITE MAP](#) [HELP](#)

Diagnostic - DSL Line

```
relative capacity occupation: 45%
noise margin upstream: 26.0 db
output power downstream: 20.0 dbm
attenuation upstream: 15.0 db
carrier load: number of bits per symbol (tone)
tone 0- 31: 00 00 00 02 22 45 66 77 77 66 65 66 54 43 33 00
tone 32- 63: 00 00 00 00 02 23 44 44 55 55 55 56 66 66 66 66
tone 64- 95: 66 66 66 62 66 66 66 66 66 66 66 66 66 66 55 55
tone 96-127: 55 55 55 55 55 55 55 55 55 55 55 55 55 54 55
tone 128-159: 55 55 55 55 55 55 55 55 55 55 55 55 55 55 55
tone 160-191: 55 55 55 55 55 54 54 44 44 44 44 44 44 44 44
tone 192-223: 44 44 44 44 44 44 43 43 43 33 43 33 33 33 33
tone 224-255: 33 33 33 33 33 33 32 22 22 22 20 00 00 00 00
```

- Wizard Setup
- Advanced Setup
- Maintenance**
 - System Status
 - DHCP Table
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 - Firmware
- Logout

SITE MAP HELP

Diagnostic - DSL Line

```

relative capacity occupation: 41%
noise margin downstream: 30.5 db
output power upstream: 12.0 dbm
attenuation downstream: 32.0 db
carrier load: number of bits per symbol(tone)
tone 0- 31: 00 00 00 02 22 45 66 77 77 66 65 66 54 43 33 00
tone 32- 63: 00 00 00 00 02 23 44 44 55 55 55 56 66 66 66 66
tone 64- 95: 66 66 66 62 66 66 66 66 66 66 66 66 66 66 66 55 55
tone 96-127: 55 55 55 55 55 55 55 55 55 55 55 55 55 55 54 55
tone 128-159: 55 55 55 55 55 55 55 55 55 55 55 55 55 55 55 55
tone 160-191: 55 55 55 55 55 54 54 44 44 44 44 44 44 44 44 44
tone 192-223: 44 44 44 44 44 44 43 43 43 33 43 33 33 33 33 33
tone 224-255: 33 33 33 33 33 33 32 22 22 22 20 00 00 00 00 00

```